

Design and implementation of optimized parallel prefix based FIR filters for software defined radiochannelizer applications.

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Abstract— For the software defined applications, the different approaches to solve the channel equalize problems have been presented. If the conventional multipliers and adders are used it tends to delay the overall system to rectify the mistake reconfigurable multipliers, adders and other logic designs. We have tried to optimize the different parameters like area, delay, power leakage, throughput and latency. Verilog HDL is used in this design. The newly designed distributed arithmetic based reconfigurable FIR is designed 64 tap filter length. The comparison between different parameters has been carried out. The results of the existing and proposed architecture is done. Throughput and latency have improved by 14.3% and 23%.

Keywords— Distributed arithmetic, LUT, Finite Impulse response, Residual number system, Software Defined radio.

I. INTRODUCTION

Earlier, modulo LUT based multipliers were used for DSP applications, where this multiplier is usually recommended for lower word lengths, such as 4 and 8 bits, which take up a lot of space. To multiply the larger numbers, in proposed modulo $2N-1$ multiplier which is memory less and can be realized by using binary adders, multipliers and logic gates. When various outcomes are processed in equal amounts, the FIR channel plan also depends on the number of delay units used, the capacity of coefficients, and the record of fractional values in multipliers. With adjustable and longer channel lengths and increased use of registers free of square size, DA provides a superior solution to the aforementioned challenges. For the extraction of restricted-band signals, a channel equalizer is required. The usage of a FIR channel allows for better narrow band signal extraction from channels. The proposed FIR channel has 64 taps, each of which is split into four LUTs, each of which holds 16 channel coefficients. One of the most difficult tasks in computerized correspondence connected to Software Defined Radio is extracting narrow channels from wide band channels. The extraction of narrow band signal is typically the testing period in remote correspondence. To enable

advanced correspondence for trans-collector of multipurpose, various information transmission, SDR completely relies on restricted band signal. Optimization of FIR filter can be achieved by using parallel prefix adder based FIR filters. The overall effect of noise, delay on system can be reduced by applying DA based parallel prefix adder filters.

Power consumption in normal filter is more, instead we can use the parallel prefix based adders and filters.

High-speed is obtained by introducing the high speed DA based FIR filters.

II. PROBLEM STATEMENT

If we go for any type of filters the major concern is about 1. Power consumption.

1. Delay.

Power consumption happens mainly due to high transitions happening in the processing of data.

If we are able to develop an algorithm to reduce the number of transitions we will be able to reduce the power. Proposed algorithm. The innovative Distributed Arithmetic-Residue Number System (DA-RNS)-based architecture for high-speed Finite Impulse Response channel. In the proposed technique, information is handled in Binary Code (BC), while the result is represented using Look-up Tables (LUTs). The usage of LUTs is advantageous since it avoids reasoning complexity caused to the $2k$ modulo factor.

III. Proposed Algorithm

The proposed PPA gathers the LUT result in a short term basis, and productivity can be improved by 12%. The number of adders required for the

equalisers application is 816 . The main segment in Fig 1 is perusing of LUT esteems in equal and exposed to computational activity , the outputs of midway products are taken care of to PPA to build the item outputs with little delay of PPA inactivity, as shown. The coefficients stored in the LUTs are shared equally among all of the registers, resulting in (x-k) information testing. When compared to traditional dLUT, the sharing concept significantly reduces delay and region utilisation.

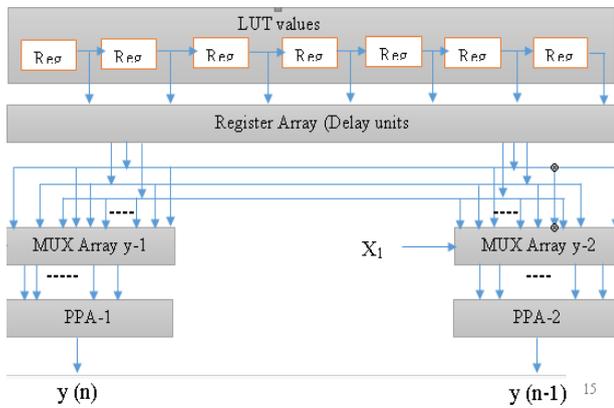


Fig 1.Optimized LUT in DA RFIR filter design

A. Partial Products Generator and Coefficients of filter

Fig.1shows the concurrent DA-based RFIR filter for N=64, number of LUTs size L=4 and the number of coefficients in each LUT is 16. LUT's and Parallel Prefix Adder (PPA). In Shift Adder Tree (SAT) proposed for LUT coefficients storage and to add partial products which are generated in multiplier RFIR for optimization of delay Then number of adder required in SAT for N=64 is 2072 .Further optimization can bedone in terms of delay and number adder by incorporating parallel prefix adder to the DA based multiplieIn the brief time frame energy-based VAD, the energy of the discourse signal casing is contrasted and the edge contingent upon the commotion level.

The proposed PPA gathers the LUT values to parallel waydeferralandproductivitycan be improved by 12% .For SDR equalizer application number of addersrequired is 816 .

The channelcoefficientsstores in 4 LUT's forplaying out the RFIR channelactivities and applied forSDR channel equalizer to deliver commotion less y(n) and changed engineering of PPGC store is displayed in Fig.3.4..All Coefficients are

putaway in all LUT's and these Coefficients are added with PPGC yields to createnumber of incomplete items and these are added utilizing PPA .

Each LUT havechannel selectors as addresses to choose the necessary .The refreshed LUTCoefficients exhaustive M MUX cluster are gets the information by fractional item and contribution to PPA.

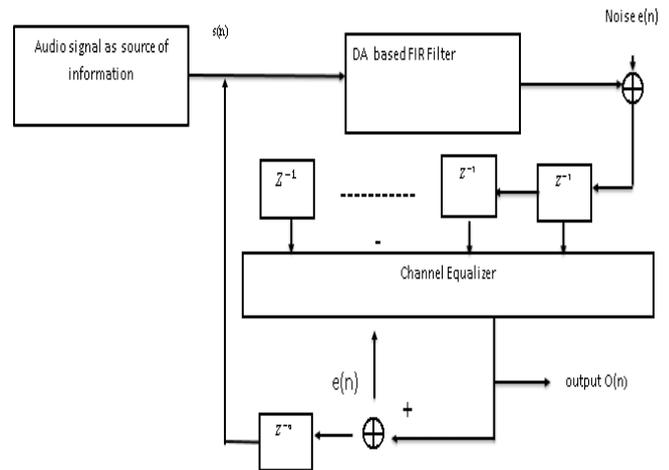


Fig.2.Block diagram of Channel Equalizer of SDR application applied to DA-LUT based RFIR

IV. EXPERIMENTAL RESULTS

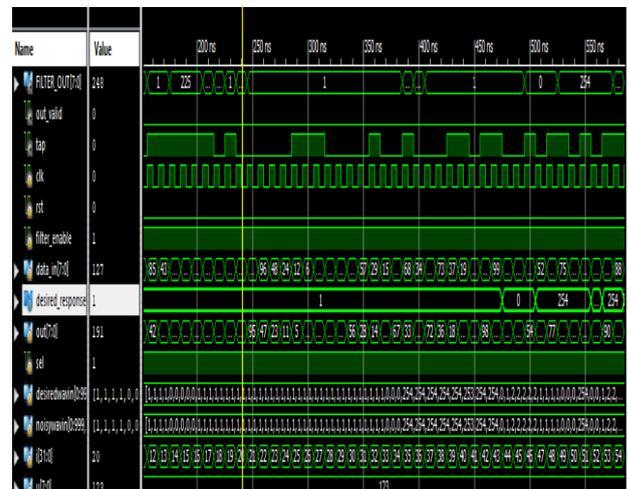


FIG 3: . Simulated results of DA-LUT RFIR on Xilinx.

For the given delay units along its qualities and coefficients are registered utilizing customary strategy as given above and its last value is 96.

$$y[n] = x[n-3]*h[n-3]+x[n-2]*h[n-2]+x[n-1]*h[n-1]+x[n]*h[n]$$

$$y [n] =x[0]*h[0]+x[1]*h[1]+x[2]*h[2]+x[3]*h[3]$$

