

# HIGH STEP-UP INTERLEAVED DC-DC CONVERTER WITH PI CONTROLLER FOR RENEWABLE ENERGY APPLICATIONS

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**Abstract—** This paper proposes a novel high step-up interleaved boost converter suitable for distributed generation using renewable and alternative power sources. The proposed interleaved boost converter not only lengthens the lifetime of the renewable power source by reducing the input-current ripple but also achieves high step-up conversion. In addition, the voltage stress of the main switches is lowered due to the lossless passive-clamp

circuit. Hence, large voltage spikes across the main switches are alleviated and the efficiency is improved. Finally the proposed system offers an excellent dynamic performance for boost the DC voltage that is verified in MATLAB simulink 2017. Experimental setup is also did in scaled down version and its outputs are verified..

## I.Introduction

For overcoming energy-shortage and environmental-contamination issues, renewable and alternative power sources that feature cleanliness and sustainability play an important role in the world, and have begun to be employed worldwide for environment protection. The voltage levels of renewable and alternative power sources, such as photovoltaic cells and fuel cells, are generally low. Thus, high-step-up DC-DC converters have been widely utilized in such renewable energy systems in order to boost their voltage levels. The high step-up DC-DC converter can convert low levels of input voltage (typically 40~50V) from renewable sources into high levels of output voltages (typically 380~400V), which are then fed into a DC load or a DC-AC inverter for supplying AC sources with an AC load. Hence, the high-step-up DC-DC converter with high efficiency is essential in such power-conversion systems.

The conventional step-up converters, such as the boost converter and fly back converter, obtain high voltage gain by adopting an extremely high duty cycle or high turns ratio of the coupled inductor. The circuit efficiency of these converters is limited due to the equivalent resistances or from the leakage inductance of windings, and high voltage spikes and stresses occur on the semiconductor devices. Adopting an extremely high duty cycle results in large conduction losses, serious diode reverse-recovery problems, and electromagnetic interference (EMI) issues. Because of the high voltage stresses that occur on the power devices, power switches with low  $R_{DS(ON)}$  and power diodes with low reverse-

recovery time cannot be employed in this type of high-step-up converter.

Some existing converters that utilize coupled inductors to achieve high voltage conversion ratio, which recycle the leakage-inductance energy and lower the voltage stresses, have been proposed. Interleaved converter with built-in transformer and interleaved converter with voltage multiplier module or with coupled inductor are another superior solution to obtain high step-up voltage gain and lower input current ripple. The coupled-inductor deals with large DC magnetizing current, so the volume of core is larger and an air gap is required to avoid saturation of core; thus, the cost is higher and the efficiency is lower. On the other hand, the built-in transformer does not deal with large DC magnetizing current, and the voltage gain can be extended by increasing the turn's ratio of the built-in transformer without an air gap; thus, the volume of core is smaller and the coupling coefficient as well as the circuit efficiency is higher. This paper proposes a novel high-step-up interleaved boost converter that not only utilizes the clamp capacitors but also integrates the secondary winding of the built-in transformer; thus, high step-up voltage gain of the presented converter and lower voltage stresses of the power devices are achieved.

The proposed interleaved boost converter with features of high step-up conversion, high circuit efficiency, and low input-current ripple, which can lengthen the life time of the input source, is suitable for distributed generation using renewable and alternative power sources. In addition, windings of the built-in transformer can be designed to extend the step-up gain, and two diodes and two capacitors in

the proposed converter act as an active clamp circuit in order to lower voltage stress on the main switches; thus, low-voltage-rated semiconductor devices (such as power MOSFETs and diodes) can be adopted in the presented converter.

The key characteristics of the proposed converter are listed as follows: (1) Lowering the input-current ripple and reducing the conduction losses results in an increased lifetime of the power sources and makes the presented converter suitable for renewable and alternative energy applications. (2) The converter is capable of achieving high step-up gain easily. (3) By recycling the leakage energy, the voltage stresses of clamp diodes are alleviated and the circuit efficiency is improved. (4) The voltage stresses on the semiconductor components are substantially lower than the output voltage.

Compared with existing converter, the proposed high step-up converter decreases the power switch count and achieves similarly high circuit efficiency without soft-switching function and active clamp circuit. Moreover, the proposed converter has features of cost-effectiveness and relatively low input current ripple in comparison with others.

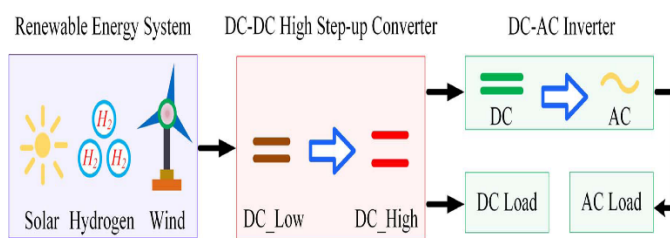
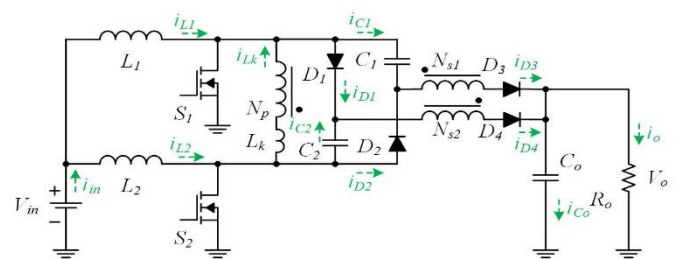


Figure 1.1: Block diagram of distributed generation using renewable power sources

## II. INTERLEAVED BOOST CONVERTER

The proposed high-step-up interleaved boost converter is shown in Fig. 4.1, where  $L_1$  and  $L_2$  are the energy storage inductors;  $S_1$  and  $S_2$  denote the power switches;  $C_1$  and  $C_2$  are the clamp capacitors;  $C_o$  is the output capacitor;  $D_1$  and  $D_2$  are the clamp diodes, and  $D_3$  and  $D_4$  are the rectified diodes. The built-in transformer consists of a primary winding  $N_p$ , a secondary winding  $N_{s1}$ , a third winding  $N_{s2}$  and a leakage inductor  $L_k$ .



The gate-driving signals of the two power switches are interleaved with a 180-degree phase shift, and the theoretical waveform of the proposed converter operating in continuous-conduction mode (CCM) is depicted in Fig.4.2.

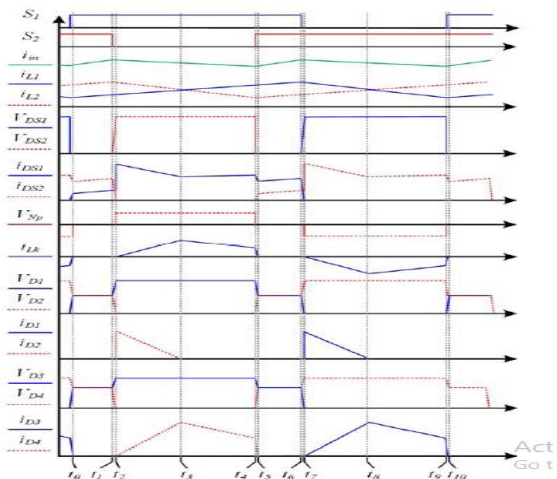


Fig. 4.3 shows the corresponding operational modes of the equivalent circuit. There are 10 main operational modes in one switching period. Due to the completely symmetrical interleaved topology,

operating modes 1 to 5 and 6 to 10 are similar. In order to simplify the analysis of the proposed converter's operating principle, only modes 1 to 5 are analyzed and discussed.

*Mode 1* [ $t_0, t_1$ ]:

At  $t=t_0$ , both power switches ( $S_1$  and  $S_2$ ) turn on. All the diodes ( $D_1, D_2, D_3$ , and  $D_4$ ) are reverse-biased. The path of current flow is shown in Fig. 4(a). Inductors  $L_1$  and  $L_2$  are charged by input voltage  $V_{in}$ , and currents through inductors  $L_1$  and  $L_2$  linearly increase

*Mode 2* [ $t_1, t_2$ ]:

At  $t=t_1$ , power switch  $S_2$  turns off, and its parasitic capacitor is charged by inductor current  $i_{L2}$ . The path of current flow is shown in Fig. 4(b).

*Mode 3* [ $t_2, t_3$ ]:

At  $t=t_2$ , power switch  $S_2$  remains off. The voltages of clamp diode  $D_2$  and rectified diode  $D_4$  decrease; then  $D_2$  and  $D_4$  begin to turn on at  $t=t_2$ . The path of current flow is shown in Fig. 4(c). The input voltage  $V_{in}$  and inductor  $L_2$  provide energy to leakage inductor  $L_k$  and primary winding  $N_p$  through switch  $S_1$ , and to clamp capacitor  $C_1$  through  $S_1$  and  $D_2$ . The drain-source voltage of power switch  $S_2$  is clamped by capacitor  $C_1$ . The input voltage  $V_{in}$ , inductor  $L_2$ , capacitor  $C_2$  and secondary winding  $N_{s2}$  provide energy to output capacitor  $C_o$  and to load  $R_o$  through  $D_4$ .

*Mode 4* [ $t_3, t_4$ ]:

At  $t=t_3$ , power switch  $S_2$  is still off. The diode current  $i_{D2}$  decreases to zero and the clamp capacitor voltage

$V_{C1}$  is equal to the drain-source voltage of power switch  $S_2$ . The path of current flow is shown in Fig. 4(d). The rectified diode current  $i_{D4}$  is proportional to leakage-inductor current  $i_{Lk}$ .

*Mode 5* [ $t_4, t_5$ ]:

At  $t=t_4$ , the power switch  $S_2$  turns on. The rectified diode  $D_4$  remains forward-biased because leakage inductor current  $i_{Lk}$  still exists. The path of current flow is shown in Fig. 4(e). The leakage inductor current  $i_{Lk}$  decreases to zero at  $t=t_5$ , and rectified diode  $D_4$  begins to be reverse-biased.

In addition, the passive lossless clamp circuit is composed of two capacitors ( $C_1$  and  $C_2$ ) and two diodes ( $D_1$  and  $D_2$ ). Referring to *Mode 3* shown in Fig. 4(c), the input voltage  $V_{in}$  and inductor  $L_2$  provide energy to the clamp capacitor  $C_1$  through  $S_1$  and  $D_2$ . Thus, the drain-source voltage of power switch  $S_2$  is clamped by capacitor  $C_1$ . Referring to *Mode 8* shown in Fig. 4(h), the input voltage  $V_{in}$  and inductor  $L_1$  provide energy to the clamp capacitor  $C_2$  through  $S_2$  and  $D_1$ . Therefore, the drain-source voltage of power switch  $S_1$  is clamped by capacitor  $C_2$ .

## 4.2. VOLTAGE STRESS AND VOLTAGE STRAIN

The voltage on clamp capacitors  $C_1$  and  $C_2$  can be expressed as

$$V_{C1} = V_{C1} = \frac{1}{1-D} V_{in}$$

The voltage on output capacitor  $C_0$  can be derived

from

$$V_{C0} = \frac{2+n}{1-D} V_{in}$$

The output voltage  $V_0$  is given by

$$V_0 = V_{C0} = \frac{2+n}{1-D} V_{in}$$

In addition, the voltage gain of the proposed converter is described as

$$\frac{V_0}{V_{in}} = \frac{2+n}{1-D}$$

The voltage stresses on power switches  $S_1$  and  $S_2$  are clamped, and are derived from

$$V_{DS1} = V_{DS2} = \frac{1}{1-D} V_{in}$$

The voltage stress on diodes  $D_1$ ,  $D_2$ ,  $D_3$  and  $D_4$  are respectively given by

$$V_{D1} = V_{D2} = \frac{2}{1-D} V_{in}$$

And

$$V_{D3} = V_{D4} = V_0 = \frac{2+n}{1-D} V_{in}$$

### 2.3 DESIGN CONSIDERATIONS

In the proposed high step-up interleaved boost converter, the input current  $I_{in}$  and the ripple current  $\Delta i_L$  of the inductor are represented by

$$I_{in} = \frac{2+n}{1-D} I_0 = \left(\frac{2+n}{1-D}\right)^2 \frac{V_{in}}{R_0}$$

And

$$\Delta I_L = \frac{V_{in} \cdot D}{f_s \cdot L_1} = \frac{V_{in} \cdot D}{f_s \cdot L_2}$$

The relationship between input current  $I_{in}$  and

the ripple current  $\Delta i_L$  of the inductor in

boundary-conduction mode (BCM) is given by

$$\frac{I_{in}}{2} = \frac{\Delta I_L}{2}$$

Substituting all the equations, the boundary condition for the normalized inductor time constant, which is represented by  $\tau_{LB}$ , is

expressed by

$$\tau_{LB} = \frac{L_1}{R_0} \cdot f_s = \frac{L_2}{R_0} \cdot f_s = \frac{D \cdot (1-D)^2}{(2+n)^2}$$

Fig.4.4 shows the relationship between the boundary condition for the normalized inductor time constant  $\tau_{LB}$  and duty cycle  $D$  under a turns-ratio  $n$  of 1 according to equation, and this figure is a design guideline for selecting appropriate inductors  $L_1$  and  $L_2$  of the presented converter.

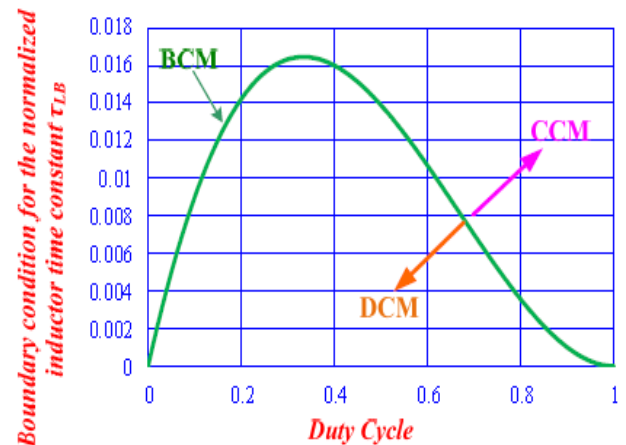


Fig.4.4. The relationship between boundary condition for the normalized inductor time constant  $\tau_{LB}$  and duty cycle



### III. SIMULATION INTERLEAVED BOOST CONVERTER

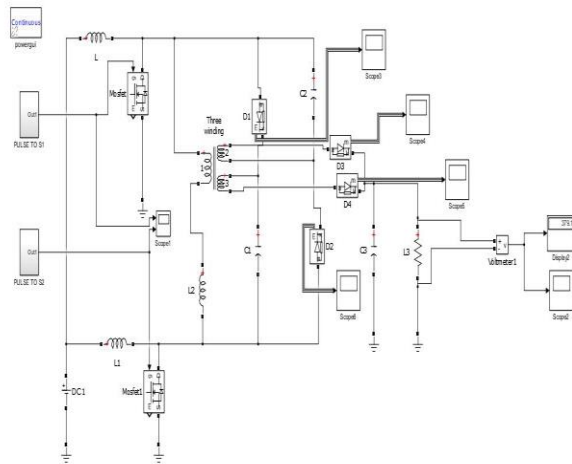


Fig.3.1. Simulation model of the interleaved boost converter

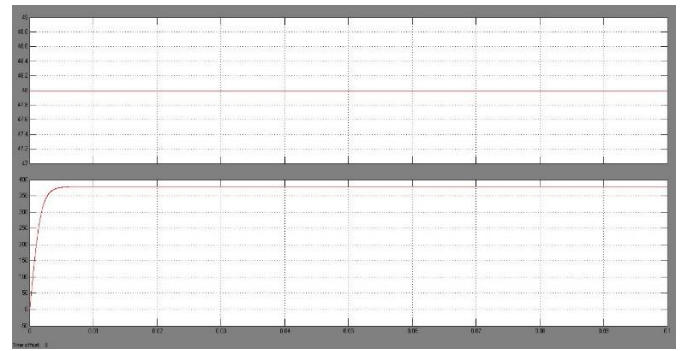


Fig 3.3 Input and output voltage of interleaved boost converter

### IV Hardware implementation and results

#### 7.1. HARDWARE DESCRIPTIONS

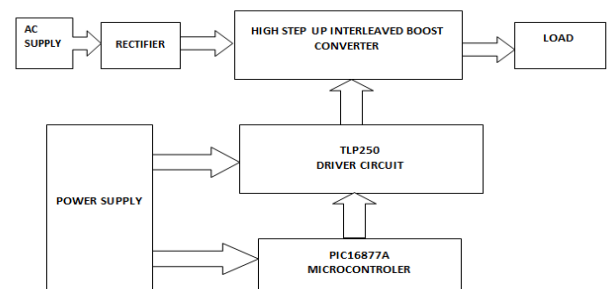


Fig 4.1 Block Diagram of Experimental Set up

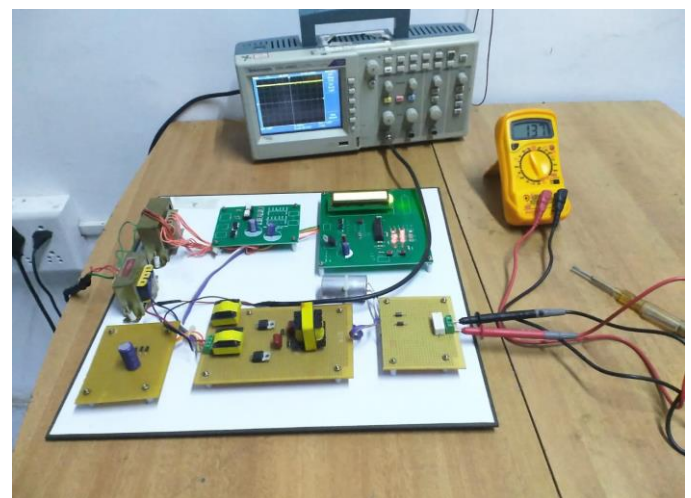


Fig. 4.2 Hardware

#### 3.2. VOLTAGE STRESS COMPARISON

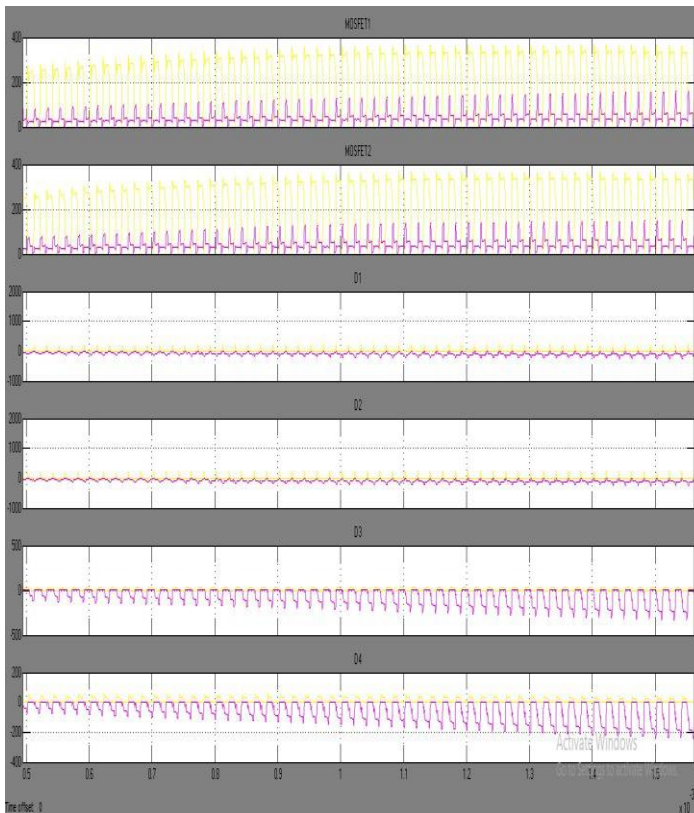


Fig 3.2. Voltage stress comparison of the proposed interleaved boost converter

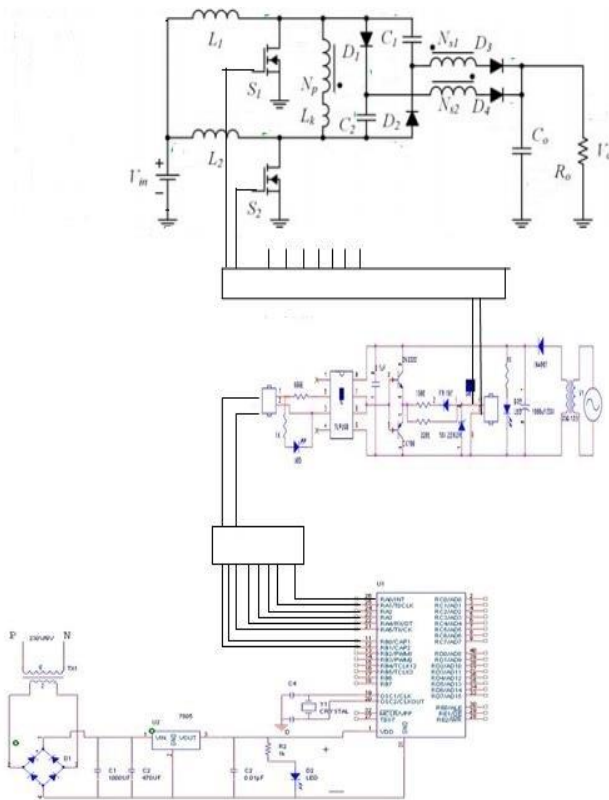


Fig 4.3 Circuit diagram

The conventional step-up converters, such as the boost converter and fly back converter, obtain high voltage gain by adopting an extremely high duty cycle or high turns ratio of the coupled inductor. The circuit efficiency of these converters is limited due to the equivalent resistances or from the leakage inductance of windings, and high voltage spikes and stresses occur on the semiconductor devices. Adopting an extremely high duty cycle results in large conduction losses, serious diode reverse-recovery problems, and electromagnetic interference

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## V Conclusion

This study has proposed a high-step-up interleaved boost converter for distributed generation using renewable and alternative power sources. The analysis of operational modes, voltage gain, and stresses is provided, and a scale down version of prototype converter has been developed and tested. The interleaved structure inside the presented converter reduces the input current ripple and distributes the current through each component. In addition, the lossless passive-clamp circuit recycles the leakage energy and constrains voltage spikes across power switches. Consequently, satisfactory experimental results have demonstrated the functionality of the proposed converter with the advantages of high step-up voltage gain, high efficiency, and suitability for Renewable and alternative energy applications. The hardware is implemented at lower power level but we can extend the implementation of this converter to high power level. By using renewable energy source with this

converter, this methodology can be used for energy harvesting.

## VI Reference

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